

ANDI-SERVO



Dual Servo Motor Controller PC/104 Card

Technical Reference Manual

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TABLE OF CONTENTS

1 General 5

 1.1 Building a System 6

2 HARDWARE..... 7

 2.1 Base Address Selection..... 7

 2.2 Interrupt Line Selection..... 8

 2.3 Motor Connector P3..... 9

 2.4 Encoder connector P1 11

 2.5 Auxiliary PWM-output connector P2..... 12

 2.6 Fault LED..... 13

 2.7 Technical specifications 13

 2.8 Board Layout 14

 2.9 Connector Locations..... 15

 2.10 Block Diagram 16

3 FUNCTIONAL DESCRIPTION 17

 3.1 General Description..... 17

 3.2 Features 17

 3.3 Theory of Operation..... 18

 3.3.1 Introduction..... 18

 3.3.2 Position feedback interface 18

 3.3.3 Velocity profile (trajectory) generation..... 19

 3.3.4 PID compensation filter..... 21

 3.4 Typical Applications..... 23

 3.4.1 Programming LM629 Host Handshaking (Interrupts) 23

 3.4.2 Incremental Encoder Interface..... 23

 3.4.3 I/O port usage (I/O map) 24

4 SOFTWARE FUNCTIONS 25

CHAPTER 1

INTRODUCTION

1 General

ANDI-SERVO is a two axis DC-servo motor-controller board for closed loop positioning applications. The board features a motor-controller processor and a H-bridge power output stage for each axis.

Controlling and driving a DC-servomechanism is a difficult task. The motor-controller is required to perform its computing in real-time and the mechanical system often involves torques and forces of a resonant nature. Mechanical systems have characteristics familiar to electronics; step response times, latency, resonance, damping etc. It is the duty of a motor-controller processor to ensure that the desired motion is performed with optimum result, even under non-ideal conditions.

The motor-controller processors on ANDI-SERVO can be parametrized for optimum performance. 32-bit position, velocity and acceleration registers, a programmable PID-filter with 16-bit coefficients and a programmable derivative sampling interval are keys to a successful implementation.

ANDI-SERVO can be used in both position- and velocity-mode systems. In the first mode, The board accelerates the motor to the desired velocity, while it is constantly monitoring the feedback to see if a desired position is reached. The board stops the motor movement, smoothly or abruptly, precisely at the desired position. In the velocity mode, the board maintains a constant velocity at a motor until the board is instructed to stop.

The ANDI-SERVO is shipped with a comprehensive software library in 'C' with over 40 functions serving as an interface to the motor-controller processor.

We recommend downloading the National Semiconductor's application notes and specifications sheets as they contain valuable information for programmers and systems integrators. The AN-693 Programming Guide and AN-706 User Guide can be downloaded from Ajeco's web site at www.ajeco.com

1.1 Building a System

The ANDI-SERVO base address must be set before it can be accessed from the host computer. The board needs 8 continuous I/O addresses, and the demo/test programs assume that base 300h can be used. Close the base jumpers 0 through 3, leave 4 open to configure the board for base 300h. The interrupt line is not needed at this stage.

The possibly easiest way to start developing a servo system, is to connect a shaft encoder and a (small energy) DC-motor to channel 1 of the board. The shaft encoder can for example be a Bourns ENS1J-B28-L00256, or Hewlett-Packard HEDS-5700 series. Any "standard" industrial incremental shaft encoders can be used. Encoders that are equipped with a separate index signal can also be used. (See sections: Motor Connector P3, Encoder Connector P1).

Apply a V_{motor} voltage to the Motor Connector P3. Observe that the PWM-output H-bridge features an undervoltage lockout ($9V_{\text{min}} \dots 11V_{\text{max}}$) at which the outputs turn OFF. Select a DC-motor that runs OK with 12...24 Vdc. Provide correspondingly a sufficient motor voltage (V_{motor}) of +12..+24 Vdc.

Run the demo program supplied on the board's software diskette. The demo program reads the shaft encoder and writes the result on the computer screen. The shaft encoder values should change on the screen, when manually rotating its axis. The DC-motor should also respond accordingly.

When the DC-motor and shaft encoder is properly connected to the board more precise and thorough software experimenting can begin.

Caution !

A typical pitfall, when building a closed loop positioning system, is that the two phase signals from the shaft encoder are interchanged. In this case, a "runoff" situation can occur. For example, the ANDI-SERVO board might drive the motor to one direction expecting that the encoder readout should decrease. If the encoders phase signals were interchanged, the encoder readout will increase instead of decrease, at which case the board will apply more force to the motor, driving it even further away from the desired position, rendering the motor with full drive applied.

CHAPTER 2

2 HARDWARE

2.1 Base Address Selection

ANDI-SERVO has five jumpers for selecting the board's base address. The jumpers are located close to the PC/104 connector, labeled BASE 0 through BASE 4. The board's base address must be set before it can be accessed from the host CPU. The board occupies 8 contiguous I/O addresses including the selected base address. (If base 300h is selected, addresses 300h through 307h will be occupied by the board.)

The list describes how the jumper selections correspond with the selected base address. When a jumper is closed (or mounted), it is referred to as a '0'. When a jumper is removed (dismounted), it is referred to as a '1'. The five jumpers enable 32 different base addresses.

Base Address					Base Address						
4	3	2	1	0		4	3	2	1	0	
0	0	0	0	0	200h	1	0	0	0	0	300h
0	0	0	0	1	210h	1	0	0	0	1	310h
0	0	0	1	0	220h	1	0	0	1	0	320h
0	0	0	1	1	230h	1	0	0	1	1	330h
0	0	1	0	0	240h	1	0	1	0	0	340h
0	0	1	0	1	250h	1	0	1	0	1	350h
0	0	1	1	0	260h	1	0	1	1	0	360h
0	0	1	1	1	270h	1	0	1	1	1	370h
0	1	0	0	0	280h	1	1	0	0	0	380h
0	1	0	0	1	290h	1	1	0	0	1	390h
0	1	0	1	0	2A0h	1	1	0	1	0	3A0h
0	1	0	1	1	2B0h	1	1	0	1	1	3B0h
0	1	1	0	0	2C0h	1	1	1	0	0	3C0h
0	1	1	0	1	2D0h	1	1	1	0	1	3D0h
0	1	1	1	0	2E0h	1	1	1	1	0	3E0h
0	1	1	1	1	2F0h	1	1	1	1	1	3F0h

Table A. Base Address of Jumpers 4 - 0

2.2 Interrupt Line Selection

ANDI-SERVO can be jumpered to use interrupt lines 2,3,4,5,6 or 7. The interrupt line is selected by mounting a jumper block at the corresponding pins on the board. The interrupt selection lines are labeled on the silk screen.

The board can generate interrupts from four sources: Both of the two H-bridge output stages and both of the motor-controller processors can cause an interrupt. The interrupt lines are "OR"-ed on the board.

The interrupt source can be determined by reading the interrupt "cause" register at *BASE+6*. The H-bridge output stage interrupts are purely diagnostic - the interrupt is generated at an overheat condition. If the output stage is overheated (145 °C), it will cause an interrupt. The motor-controller processor is capable of generating interrupts from a multitude of (programmable) sources.

When read, the *BASE+6* register's bits are decoded as follows:

- Bit-0 If '1', Thermal flag interrupt from H-bridge output stage, channel-1
- Bit-1 If '1', Thermal flag interrupt from H-bridge output stage, channel-2
- Bit-2 If '1', Motor-controller processor at channel-1 interrupts
- Bit-3 If '1', Motor-controller processor at channel-2 interrupts

2.3 Motor Connector P3

DC-, and brushless DC-motors can be connected to the ANDI-SERVO motor connector, labeled with the text P3 on the board's silk screen. The motor output method is 8-bit sign/magnitude PWM.

The motor connector is an 8-pin field wiring connector with 5 mm pitch. The pins are numbered from 1 through 8 having the following use:

Connector P3

Pin-1	Channel 2,	PWM-drive output A
Pin-2	Channel 2,	Vmotor, +12...50 Vdc input
Pin-3	Channel 2,	GND (not isolated from computer, see Caution !)
Pin-4	Channel 2,	PWM-drive output B
Pin-5	Channel 1,	PWM-drive output A
Pin-6	Channel 1,	Vmotor, +12...50 Vdc input
Pin-7	Channel 1,	GND (not isolated from computer, see Caution !)
Pin-8	Channel 1,	PWM-drive output B

Table 2. Connector P3

Caution !

Pay attention to potentially high motor voltages ! The absolute maximum voltage for the PWM-stage is 55 Vdc. Exceeding this level can lead to damage to the computer system and the ANDI-SERVO card. Carefully check the pinout and voltages before connecting to the board.

Pay attention that ANDI-SERVO is not isolated from the computer system. The terminals marked with: GND, MUST be connected from the P3-connector to the system's power supply ground, by own wires. This ensures that the motor ground currents will be properly conducted to the system ground by NOT going through the PC/104 stack !

The terminals marked with PWM-drive output A (and B) should be connected to the DC-motor's positive and negative terminals. Ensure that the DC-motor power terminals are not connected to system ground. (Use a DC-motor which has a direction of rotation that can be reversed by changing the polarity at its terminals).

Warning !

Never connect an oscilloscope directly over the PWM-drive output A and B terminals. The probe-ground can accidentally short the PWM-output. A proper way of measuring at the PWM-outputs, is to use two probes, both grounded to GND. Measure with one probe at output A and one at output B. The PWM-output stage is a H-bridge. The output A will be pulled to Vmotor and output B pulled to ground or vice versa, depending on the drive direction.

2.4 Encoder connector P1

Industry standard incremental shaft encoders may be used with ANDI-SERVO for positioning feedback. ANDI-SERVO has also support for encoders with index pulse. Examples of encoders are Bourns ENS1J-B28-L00xxx and Hewlett-Packard HEDS-5700 series.

The minimum encoder pulse width (high or low) is 2 μ S. The dwell time (phase-A low from phase-B high, or phase-B high from phase-A low) minimum is 1 μ S. The maximum encoder capture rate is 1 MHz.

Connector P1 Pinout:

Pin-1	GND (Not isolated from computer)
Pin-2	Phase-B, channel 2
Pin-3	Phase-A, channel 2
Pin-4	Index, channel 2
Pin-5	Phase-B, channel 1
Pin-6	Phase-A, channel 1
Pin-7	Index, channel 1
Pin-8	Vcc output (system +5 Vdc)

Table 3. Connector P1 Pinout

Note !

The lines 2, 3, 4, 5, 6 and 7 are pulled up to system +5Vdc with 1 kOhm. The encoder must be capable of sinking 5 mA.

Warning !

Use only encoders with "Open Collector" type of output. Encoder Vin-max = System Vcc.

2.5 Auxiliary PWM-output connector P2

If the motor driver circuitry on the ANDI-SERVO is not capable of sourcing enough current to the motor, it is possible to utilize the motor-controller processor part of the board and connect an external, and more powerful H-bridge to the P2 connector. This connector outputs 8-bit sign/magnitude PWM of TTL-level.

PWM-output connector P2:

Pin-1	GND (not isolated from system)
Pin-2	Thermal flag input, channel 1
Pin-3	Thermal flag input, channel 2
Pin-4	Brake output, channel 2
Pin-5	Brake output, channel 1
Pin-6	Direction output, channel 2
Pin-7	PWM-output, channel 2
Pin-8	Direction output, channel 1
Pin-9	PWM-output, channel 1
Pin-10	GND (not isolated from system)

Table 4. PWM-output connector P2

Caution !

All signals are TTL-level.

Warning !

The thermal flag inputs are pulled to Vcc (system +5Vdc) by 1 kOhm. An "open collector" type of device capable of sinking 5 mA is the only allowed type of signal source.

2.6 Fault LED

The fault LED on the ANDI-SERVO serves many purposes. It can act as an visual alarm signal or as a diagnostic tool. The LED will light up upon a thermal overheat condition, interrupt or a software call to the function `s_fault_led()`;

2.7 Technical specifications

<i>Size:</i>	91 x 96 mm (3.6" x 3.8"), PC/104
<i>Operating temp. range:</i>	0..70°C
<i>Weight:</i>	80 grams
<i>Position range:</i>	-1,073,741,824 to +1,073,741,823 counts
<i>Velocity range:</i>	0 to 16,383 counts/sample, with a resolution of 1/65536 counts/sample
<i>Acceleration range:</i>	0 to 16,383 counts/sample/sample, with a resolution of 1/65536 counts/sample/sample.
<i>Operating modes:</i>	Position and Velocity
<i>Control algorithm:</i>	PID plus programmable integration limit
<i>Max. encoder state capture rate:</i>	1 MHz (encoder signals must remain stable for 1 μ S.)
<i>PID Sample intervals:</i>	Derivative term: Programmable from 256 μ S to 65,536 μ S, in steps of 256 μ S/step. Proportional and integral term: 256 μ S.
<i>PWM-output stage:</i>	0.5 A @ 24 Vdc
<i>Absolute maximum ratings:</i>	Max. Vmotor 55 Vdc, Max. 3 A peak
<i>Thermal shutdown:</i>	@175° C
<i>Thermal warning interrupt:</i>	@145° C
<i>Undervoltage lockout:</i>	9 Vmin, 11 Vmax
<i>Encoder wire thickness:</i>	0.08..0.5mm ²

2.8 Board Layout

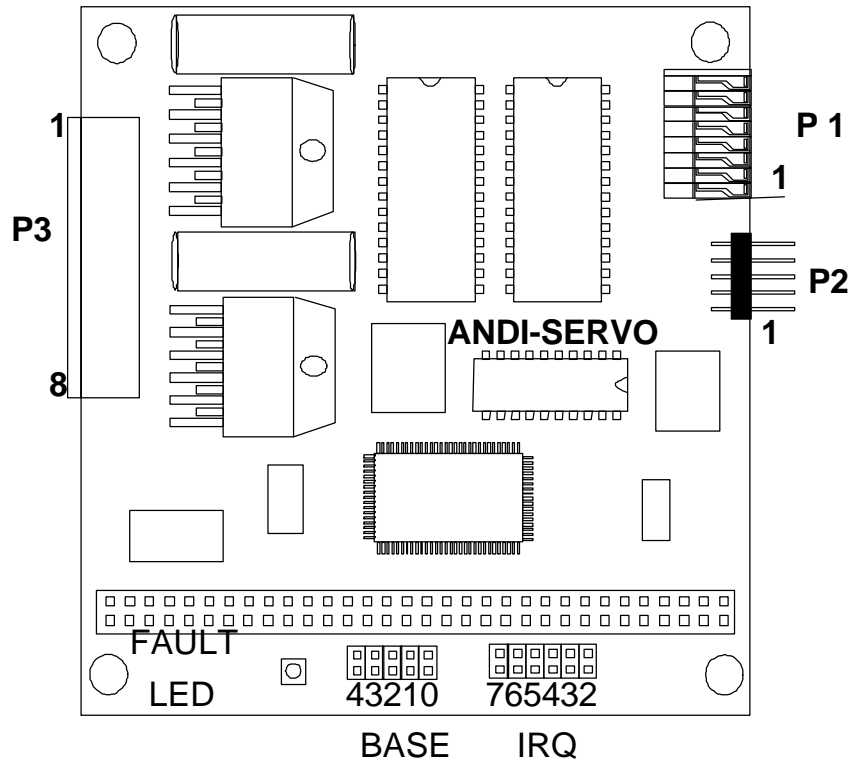


Figure 1. ANDI-SERVO Board Layout

2.9 Connector Locations

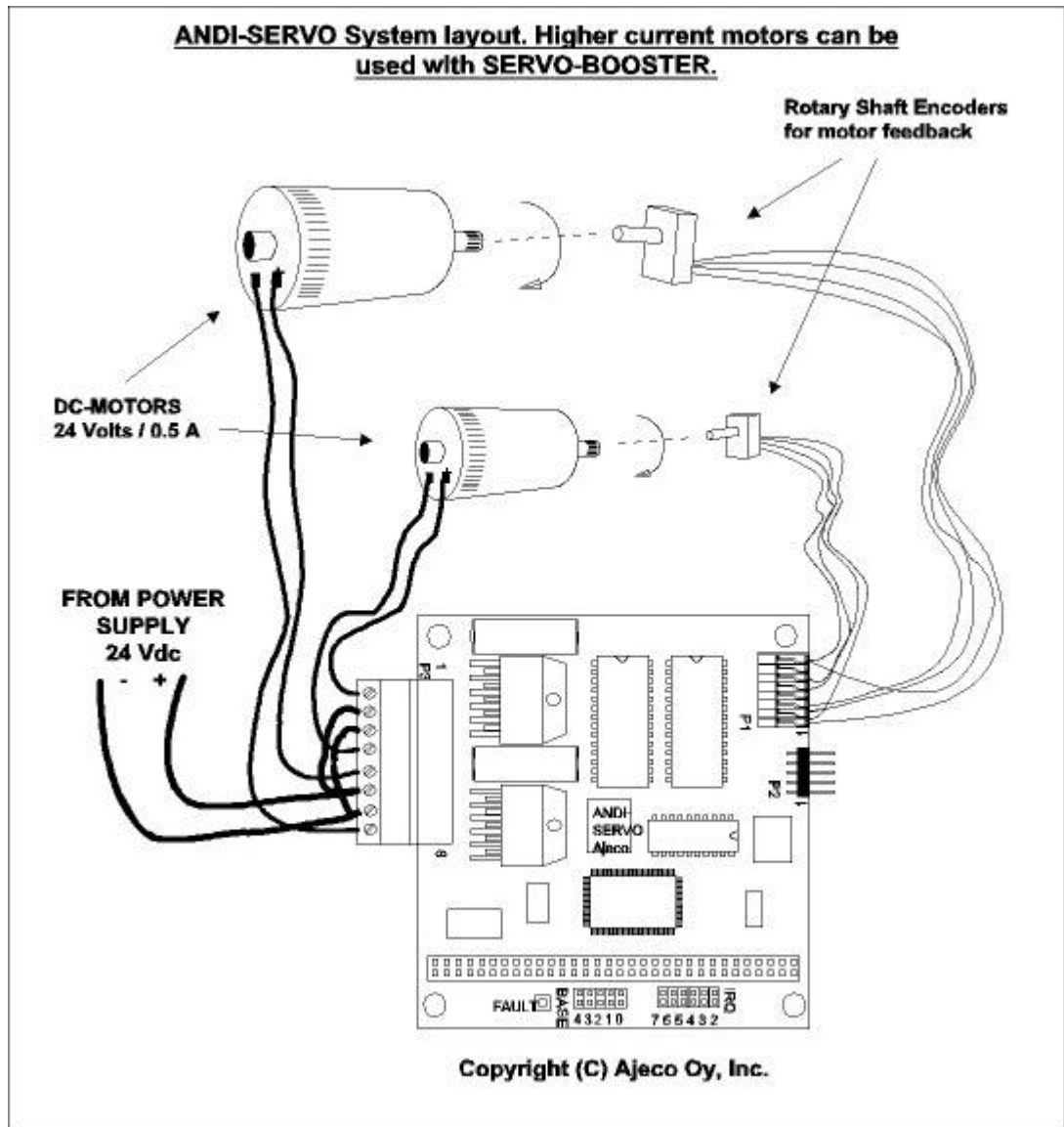


Figure 2. ANDI-SERVO connector locations

2.10 Block Diagram

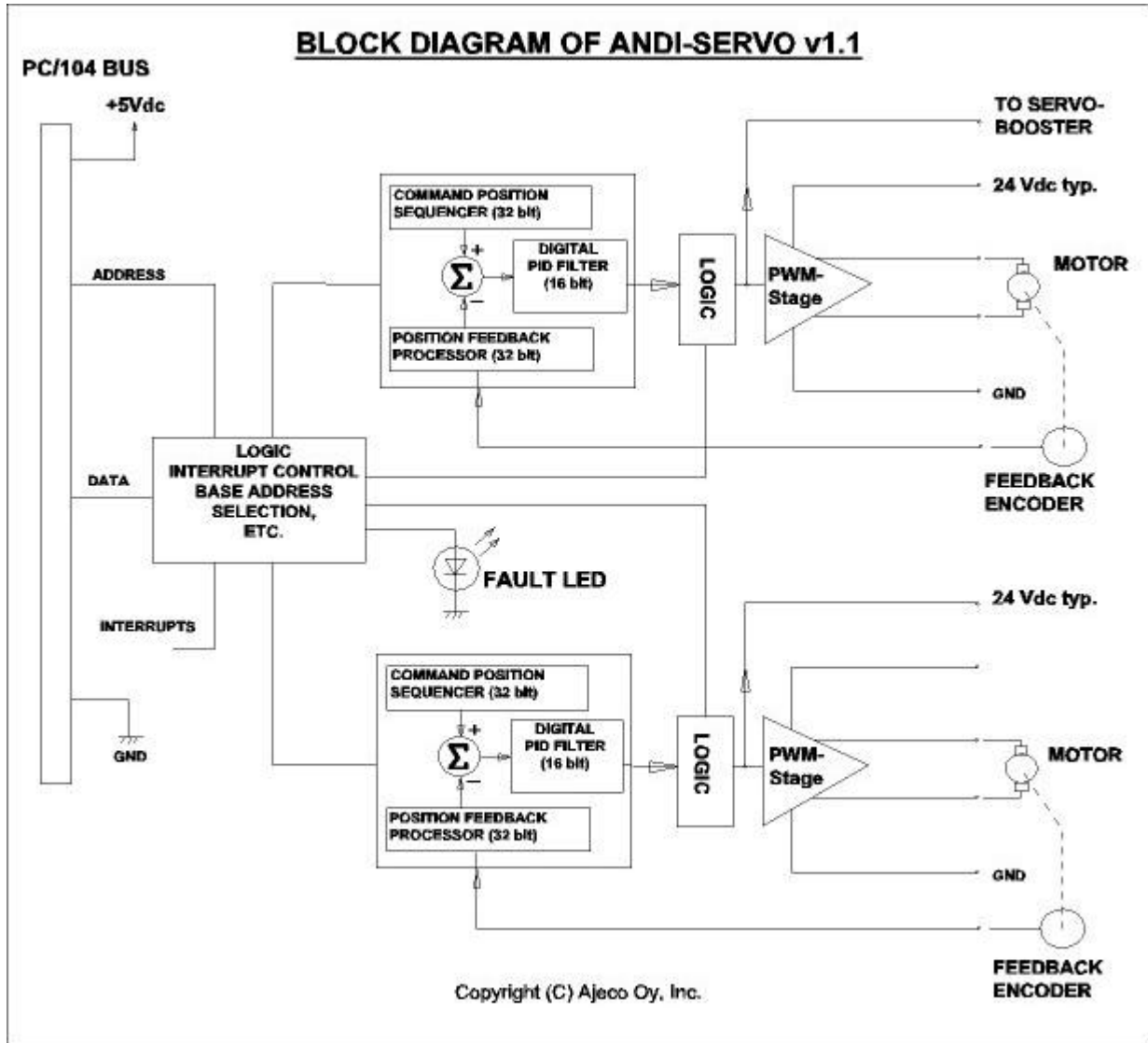


Figure 3. Block Diagram of ANDI-SERVO v1.1

CHAPTER 3

3 FUNCTIONAL DESCRIPTION

LM629 PRECISION MOTION CONTROLLER

3.1 General Description

The LM629 is a dedicated motion-control processor designed for use with a variety of DC and brushless DC servo motors, and other servomechanisms which provide a quadrature incremental position feedback signal. The part perform the intensive, real-time computational tasks required for high performance digital motion control. The host control software interface is facilitated by a high-level command set in the ANDI-SERVO library.

3.2 Features

- 32-bit position, velocity, and acceleration registers
- Programmable digital PID filter with 16-bit coefficients
- Programmable derivative sampling interval
- Internal trapezoidal velocity profile generator
- Velocity, target position, and filter parameters may be changed during motion
- Position and velocity modes of operation
- Real-time programmable host interrupts
- Quadrature incremental encoder interface with index pulse input

3.3 Theory of Operation

3.3.1 Introduction

The typical system block diagram (See *Figure 3*, ANDI-SERVO Block Diagram) illustrates a servo system built using the LM629. The host processor communicates with the LM629 through an I/O port to facilitate programming a trapezoidal velocity profile on a digital compensation filter. The PWM output interfaces to an external PWM Hbridge that produces the signal that drives the motor. An incremental encoder provides feedback for closing the position servo loop. The trapezoidal velocity profile generator calculates the required trajectory for either position or velocity mode of operation. In operation, the LM629 subtracts the actual position (feedback position) from the desired position (profile generator position), and the resulting position error is processed by the digital filter to drive the motor to the desired position. Table I provides a brief summary of specifications offered by the LM629:

3.3.2 Position feedback interface

The LM629 interfaces to a motor via an incremental encoder. Three inputs are provided: two quadrature signal inputs, and an index pulse input. The quadrature signals are used to keep track of the absolute position of the motor. Each time a logic transition occurs at one of the quadrature inputs, the LM629 internal position register is incremented or decremented accordingly. This provides four times the resolution over the number of lines provided by the encoder. Each of the encoder signal inputs is synchronized with the LM629 clock.

The optional index pulse output provided by some encoders assumes the logic-low state once per revolution. If the LM629 is so programmed by the user, it will record the absolute motor position in a dedicated register (the index register) at the time when all three encoder inputs are logic low.

If the encoder does not provide an index output, the LM629 index input can also be used to record the home position of the motor. In this case, typically, the motor will close a switch which is arranged to cause a logic-low level at the index input, and the LM629 will record motor position in the index register and alert (interrupt) the host processor. Permanently grounding the index input will cause the LM629 to malfunction.

3.3.3 Velocity profile (trajectory) generation

The trapezoidal velocity profile generator computes the desired position of the motor versus time. In the position mode of operation, the host processor specifies acceleration, maximum velocity, and final position. The LM629 uses this information to affect the move by accelerating as specified until the maximum velocity is reached or until deceleration must begin to stop at the specified final position. The deceleration rate is equal to the acceleration rate. At any time during the move the maximum velocity and/or the target position may be changed, and the motor will accelerate or decelerate accordingly.

When operating in the velocity mode, the motor accelerates to the specified velocity at the specified acceleration rate and maintains the specified velocity until commanded to stop. The velocity is maintained by advancing the desired position at a constant rate. If there are disturbances to the motion during velocity mode operation, the long-time average velocity remains constant. If the motor is unable to maintain the specified velocity (which could be caused by a locker rotor, for example), the desired position will continue to be increased, resulting in a very large position error. If this condition goes undetected, and the impeding force on the motor is subsequently released, the motor could reach a very high velocity in order to catch up to the desired position (which is still advancing as specified).

All trajectory parameters are 32-bit values. Position is a signed quantity. Acceleration and velocity are specified as 16-bit, positive-only integers having 16-bit fractions. The integer portion of velocity specifies how many counts provide increased average velocity resolution. Acceleration is treated in the same manner. With each sampling interval the commanded acceleration value is added to the current desired velocity (unless the command velocity has been reached).

One determines the trajectory parameters for a desired move as follows. If, for example, one has a 500-line shaft encoder, desired that the motor accelerates at one revolution per second until it is moving at 600 rpm, and then decelerate to a stop at a position exactly 100 revolutions from the start, one would calculate the trajectory parameters as follows:

```

let          P = target position (units = encoder counts)
let          R = encoder lines * 4 (system resolution)
then        R = 500 * 4 = 2000
and         P = 2000 * desired number of revolutions
           P = 2000 * 100 revs = 200,000 counts
           ( value to load)
           P (coding) = 00030D40 (hex code written to LM629)

let          V = velocity (units = counts/sample)
let          T = sample time (seconds) = 341 s
           (with 6 MHz clock)

let          C = conversion factor = 1 minute/60 seconds
then        V = R * T * C * desired rpm
and         V = 2000 * 341E-6 * 1/60 * 600 rpm
           V = 6.82 counts /sample
           V (scaled) = 6.82 * 65,536 = 446,955.52
           V (rounded) = 446,956 (value to load)
           V (coding) = 0006D1EC (hex written to LM629)

let          A = acceleration (units = counts/sample/sample)
then        A = R * T * T * desired acceleration (rev/sec/sec)
and         A = 2000 * 341E-6 * 341E-6 * 1 rev/sec/sec
           A = 2.33E-4 counts/sample/sample
           A (scaled) = 2.33E-4 * 65,536 = 15.24
           A (rounded) = 15 (value to load)
           A (coding) = 0000000F (hex code written to LM629)

```

The above position, velocity, and acceleration values must be converted to binary codes to be loaded into the LM629. The values shown for velocity and acceleration must be multiplied by 65,536 (as shown) to adjust for the required integer/fraction format of the input data. Note that after scaling the velocity and acceleration values, literal fractional data cannot be loaded; the data must be rounded and converted to binary. The factor of four increases in system resolution is due to the method used to decode the quadrature encoder signals.

3.3.4 PID compensation filter

The LM629 uses a digital Proportional Integral Derivative (PID) filter to compensate for the control loop. The motor is held at the desired position by applying a restoring force to the motor that is proportional to the position error, plus the integral of the error, plus the derivative of the error. The following discrete-time equation illustrates the control performed by the LM629:

$$u(n) = k_p \cdot e(n) + k_i \sum_{N=0}^n e(n) + k_d [e(n') - e(n' - 1)]$$

where $u(n)$ is the motor control signal output at sample time n , $e(n)$ is the position error at sample time n , n' indicates sampling at the derivative sampling rate, and k_p , k_i and k_d are the discrete-time filter parameters loaded by the users.

The first term, the proportional term, provides a restoring force proportional to the position error, just as does a spring obeying Hooke's law. The second term, the integration term, provides a restoring force that grows with time, and thus ensures that the static position error is zero. If there is a constant torque loading, the motor will still be able to achieve zero position error.

The third term, the derivative term, provides a force proportional to the rate of change of position error. It acts just like viscous damping in a damped spring and mass system (like a shock absorber in a automobile). The sampling interval associated with the derivative term is user-selectable; this capability enables the LM629 to control a wide range of inertial loads (system mechanical time constants) by providing a better approximation of the continuous derivative. In general, longer sampling intervals are useful for low-velocity operations.

(PID compensation filter....)

In operation, the filter algorithm receives a 16-bit error signal from the loop summing-junction. The error signal is saturated at 16 bits to ensure predictable behavior. In addition to being multiplied by filter coefficient k_p , the error signal is added to an accumulation of previous errors (to form the internal signal) and, at a rate determined by the chosen *derivative* sampling interval, the previous error is subtracted from it (to form the derivative signal). All filter multiplications are 16-bit operations; only the bottom 16 bits of the product are used.

The integral signal is maintained to 24 bits, but only the top 16 bits are used. The scaling techniques results in a more usable (less sensitive) range of coefficient k_i values. The 16 bits are right-shifted eight positions and multiplied by filter coefficient k_i to form the term which contributes to the motor control output. The absolute magnitude of this product is compared to coefficient i_l , and the lesser, appropriately signed magnitude then contributes to the motor control signal.

The derivative signal is multiplied by coefficient k_d each *derivative* sampling interval. This product contributes to the motor control output every sample interval, independent of the user-chosen *derivative* sampling interval.

The k_p , limited k_i , and k_d product terms are summed to form a 16-bit quantity. Depending on the output mode (wordsize), either the top 8 or top 12 bits become the motor control output signal.

3.4 Typical Applications

3.4.1 Programming LM629 Host Handshaking (Interrupts)

A few words regarding the LM629 host handshaking will be helpful to the system programmer. As indicated in various portions of the above text, the LM629 handshakes with the host computer in two ways: via the host interrupt output (Pin 17), or via polling the status byte for “interrupt” conditions. When the hardwired interrupt is used, the status byte is also read and parsed to determine which of six possible conditions caused the interrupt.

When using the hardwired interrupt it is very important that the host interrupt service routine does not interfere with a command sequence which might have been in progress when the interrupt occurred. If the host interrupt service routine were to issue a command to the LM629 while it is in the middle of an ongoing command sequence, the ongoing command will be aborted (which could be detrimental to the application).

Two approaches exist for avoiding this problem. If one is using hardwired interrupts, they should be disabled at the host prior to issuing any LM629 command sequence, and re-enabled after each command sequence. The second approach is to avoid hardwired interrupts and poll the LM629 status byte for “interrupt” status. The status byte always reflects the interrupt-condition status, independent of whether or not the interrupts have been masked.

3.4.2 Incremental Encoder Interface

The incremental (position feedback) encoder interface consists of three lines: Phase A, Phase B, and Index. The index pulse output is not available on some encoders. The LM629 will work with both encoder types.

3.4.3 I/O port usage (I/O map)

ANDI-SERVO occupies 8 contiguous I/O addresses, base+0 through base+7. The I/O port usage is as follows:

PORT ADDRESS	FUNCTION AT READ	FUNCTION AT WRITE	NOTES
base+0	Status byte	Command byte	from/to LM629, channel 1
base+1	Data byte	Data byte	-"
base+2	Status byte	Command byte	from/to LM629, channel 2
base+3	Data byte	Data byte	-"
base+4	No op	PWM-Brakes	bit-1=chan2, bit-0=chan1
base+5	No op	Hardware Reset	bit-1=chan2, bit-0=chan1
base+6	IRQ-cause	Ena/Dis IRQ, LED	See notes *)
base+7	Clear IRQ	No op	See notes **)

***) When written to:** This register is used for hardware enabling/disabling the interrupts and controlling the FAULT LED. If bit-1 is set to '1', interrupts are enabled, otherwise disabled. If bit-0 is set to '1', the LED will be ON, otherwise OFF.

When read: This register's lowest four bytes indicate the source of the interrupt. The register can also be occasionally polled, to check for an overheat condition, for example. The register decodes as follows:

Bit-3 = '1' = Interrupt from LM629, channel 2
 Bit-2 = '1' = Interrupt from LM629, channel 1
 Bit-1 = '1' = Thermal overheat interrupt from channel 2
 Bit-0 = '1' = Thermal overheat interrupt from channel 1

****)** The interrupt is held high until the Clear-IRQ register is read. This register does not return a value, it clears the interrupt condition.

4 SOFTWARE FUNCTIONS

Please refer to the attached software diskette for the latest library and example code. The functions herein are listed only to give a quick perception on what's included in the library code.

```
s_check_busy(int chip)
s_clr_irq(int chip, int mask)
s_disable_irq(void)
s_enable_irq(void)
s_fault_led(int mode)
s_get_base(void)
s_get_bp_abs(int chip, long *bp)
s_get_bp_rel(int chip, long *bp)
s_get_brake(int chip, int *mode)
s_get_desiredpos(int chip, long *pos)
s_get_desiredvel(int chip, long *vel)
s_get_errcode(void)
s_get_indexpos(int chip, long *pos)
s_get_integrsum(int chip, int *sum)
s_get_irqmask(int chip, int *mask)
s_get_poserr(int chip, int *trsh)
s_get_realpos(int chip, long *pos)
s_get_realvel(int chip, long *vel)
s_get_signals(int chip, uint *s)
s_get_status(int chip, int *stat)
s_get_stoperr(int chip, int *trsh)
s_init(int chip)
s_pgm_filter(int chip, S_FILTERfl)
s_pgm_trajectory(int chip, S_TRY tr)
s_reset(int chip, int mode)
s_set_base(int base)
s_set_absacc(int chip, long acc)
s_set_abspos(int chip, long pos)
s_set_absvel(int chip, long vel)
s_set_bp_abs(int chip, long bp)
s_set_bp_rel(int chip, long bp)
s_set_brake(int chip,int mode)
s_set_homepos(int chip)
s_set_indexpos(int chip)
s_set_irqmask(int chip, int mask)
s_set_poserr(int chip, int trsh)
s_set_stoperr(int chip, int trsh)
s_soft_reset(int chip)
s_start(int chip)
s_stop(int chip, int mode)
s_update_filter(int chip)
```

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